EECS 312: Digital Integrated Circuits Midterm Exam

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Show your work. Derivations are required for credit; end results are insufficient. Closed book. No electronic mental aids, e.g., calculators.

Honor Pledge: I have neither given nor received aid in this exam.

Signature:



1. (15 total pts.) Consider the following figure.

- (a) (5 pts.) What type of logic gate is this?
- (b) (10 pts.) Write an expression for the total gate output node drain capacitance as a function of λ and constants that can be found in Section 1. Note that the transistors are all minimal-length, i.e., $L = 2\lambda$. You needn't do numerical calculation. You just need to

write the expression in terms of the relevant variables. Note, V_{CC} is the system's high voltage. To those of you who already know about the Miller effect, you need not consider it in this problem. Unless you are very good at measuring, you had better show your work, e.g., by indicating your length estimates before computing capacitances based on them.

2. (10 total pts.) Consider the following equation for some sort of power consumption in an integrated circuit:

$$P_? = C \cdot V_{DD}^2 \cdot f \cdot A$$

- (a) (2 pts.) Use at most two sentences to explain what type of power consumption this equation models.
- (b) (4 pts.) Use at most one sentence to define C in this equation. A single-word definition would be insufficient.
- (c) (4 pts.) Use at most two sentences to describe how C might be reduced.
- 3. (10 total pts.) Consider the following transfer curve.



- (a) (2 pts.) What type of device is this?
- (b) (4 pts.) What are its NM_H and NM_L ?
- (c) (4 pts.) What is its V_M and g (gain)?

- 4. (10 pts.) Use three or fewer sentences to describe the properties of CMOS devices that prevented their earlier widespread use? I.e., why did BJTs long remain dominant for use in digital logic?
- 5. (10 total pts.) Draw a side view of a PMOSFET.
 - (a) (4 pts.) Label each region with the type of material used.
 - (b) (3 pts.) Label the source, drain, gate, and bulk terminals.
 - (c) (3 pts.) Indicate any diodes in the structure.
- 6. (5 pts.) Use at most one sentence to describe why an extremely low V_{DD} can result in a higher energy consumption for a particular task than a higher, but still sub-threshold, V_{DD} .
- 7. (10 pts.) If an NMOSFET gate's dielectric were changed from SiO₂ to a low- κ material with half the permittivity of SiO₂, what would happen to its I_D ? You may assume that the NMOSFET is not velocity saturated, and that its $V_{GS} \geq V_{TN}$. State any assumptions and provide evidence for your answer, e.g., by writing the applicable expression for I_D .
- 8. (10 total pts.) (7 pts.) List the following interconnect fabrication steps in chronological order:
 - Etch metal.
 - Expose photoresist using mask.
 - Remove all photoresist.
 - Deposit photoresist.
 - Deposit metal everywhere.

(3 pts.) What type of metal is this process appropriate for?

1 Reference material

	C_{OX}	C_O	C_j	m_j	ϕ_b	C_{jsw}	m_{jsw}	ϕ_{bsw}
	$(\mathrm{fF}/\mathrm{\mu m}^2)$	$(\mathrm{fF}/\mathrm{\mu m})$	$(\mathrm{fF}/\mathrm{\mu m}^2)$		(V)	$(fF/\mu m)$		(V)
NMOS	6	0.31	2	0.5	0.9	0.28	0.44	0.9
PMOS	6	0.27	1.9	0.48	0.9	0.22	0.32	0.9

MODELS FOR CMOS DEVICES

	V ₇₀ (V)	$\gamma(V^{0.5})$	$V_{DSAT}(\mathbf{V})$	k' (A/V ²)	λ (V ⁻¹)		
NMOS	0.43	0.4	0.63	115 × 10 ⁻⁶	0.06		
PMOS	-0.4	-0.4	-1	-30×10^{-6}			

CMOS (0.25 µm) - Unified Model.

CMOS (0.25 μ m) – Switch Model (R_{eq})

-1

 -30×10^{-6}

-0.1

$V_{DD}(\mathbf{V})$	1	1.5	2	2.5
NMOS (kΩ)	· 35	19	15	13
PMOS (kΩ)	115	55	38	31

CMOS (0.25 µm) - BSIM Model

See Website: http://bwrc.eecs.berkeley.edu/IcBook

Name	Value
kT/q	$25.875\mathrm{mJ/C}$
NMOSFET I_S	$21.0\mathrm{pA}$
PMOSFET I_S	$41.8\mathrm{pA}$
n (for I_D calculation)	1.5

VALUES OF MATERIAL AND **PHYSICAL CONSTANTS**

Name	Symbol	Value	Units	
Room temperature	<i>7</i> .	300 (= 27°C)	к	
Boltzman constant	k	1.38×10^{-23}	J/K	
Electron charge	4	1.6×10^{-19}	С	
Thermal voltage	$\phi_r = kT/q$	26	mV (at 300 K)	
Intrinsic Carrier Concentration (Silicon)	n _i	1.5×10^{10}	cm ⁻³ (at 300 K)	
Permittivity of Si	Ø.,	1.05×10^{-12}	F/cm	
Permittivity of SiO2	S _{ON}	3.5×10^{-13}	F/cm	
Resistivity of A1	ρ	2.7×10^{-8}	Ω-m	
Resistivity of Cu	stivity of Cu PCu		Ω-m	
Magnetic permeability of vacuum (similar for SiO ₂)	μ_0	12.6×10^{-7}	Wb/Am	
Speed of light (in vacuum)	<i>c</i> ₀	30	cm/nsec	
Speed of light (in SiO ₂)	G _{ex}	15	cm/nsec	

FORMULAS AND EQUATIONS

Diode

$$I_{D} = I_{S}(e^{V_{D}/\Phi_{T}} - 1) = Q_{D}/\tau_{T}$$

$$C_{j} = \frac{C_{j0}}{(1 - V_{D}/\phi_{0})^{m}}$$

$$K_{eq} = \frac{-\phi_{0}^{m}}{(V_{high} - V_{low})(1 - m)} \times [(\phi_{0} - V_{high})^{1 - m} - (\phi_{0} - V_{low})^{1 - m}]$$

MOS Transistor

$$V_{T} = V_{T0} + \gamma (\sqrt{-2\phi_{F} + V_{SB}} - \sqrt{-2\phi_{F}})$$

$$I_{D} = \frac{k'_{n} W}{2L} (V_{GS} - V_{T})^{2} (1 + \lambda V_{DS}) \text{ (sat)}$$

$$I_{D} = v_{sat} C_{os} W \left(V_{GS} - V_{T} - \frac{V_{DSAT}}{2} \right) (1 + \lambda V_{DS}) \text{ (velocity sat)}$$

$$I_{D} = k'_{n} \frac{W}{L} \left((V_{GS} - V_{T}) V_{DS} - \frac{V_{DS}^{2}}{2} \right) \text{ (triode)}$$

$$I_{D} = I_{S} e^{\frac{V_{GS}}{nkT/q}} \left(1 - e^{-\frac{V_{DS}}{kT/q}} \right) \text{ (subthreshold)}$$

Deep Submicron MOS Unified Model

$$\begin{split} I_D &= 0 \text{ for } V_{GT} \leq 0 \\ I_D &= k' \frac{W}{L} \left(V_{GT} V_{min} - \frac{V_{min}^2}{2} \right) (1 + \lambda V_{DS}) \text{ for } V_{GT} \geq 0 \\ \text{with } V_{min} &= \min(V_{GT}, V_{DS}, V_{DSAT}) \\ \text{and } V_{GT} &= V_{GS} - V_T \end{split}$$

MOS Switch Model

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$$\begin{split} R_{eq} &= \frac{1}{2} \bigg(\frac{V_{DD}}{I_{DSAT}(1 + \lambda V_{DD})} + \frac{V_{DD}/2}{I_{DSAT}(1 + \lambda V_{DD}/2)} \bigg) \\ &\approx \frac{3}{4} \frac{V_{DD}}{I_{DSAT}} \bigg(1 - \frac{5}{6} \lambda V_{DD} \bigg) \end{split}$$

Inverter

$$\begin{split} V_{OH} &= f(V_{OL}) \\ V_{OL} &= f(V_{OH}) \\ V_{M} &= f(V_{M}) \\ t_{p} &= 0.69 R_{eq} C_{L} = \frac{C_{L}(V_{swing}/2)}{I_{avg}} \\ P_{dyn} &= C_{L} V_{DD} V_{swing} f \\ P_{stat} &= V_{DD} I_{DD} \end{split}$$

Static CMOS Inverter

$$\begin{aligned} &V_{OH} = V_{DD} \\ &V_{OL} = GND \\ &V_{M} \approx \frac{rV_{DD}}{1+r} \quad \text{with} \quad r = \frac{k_p V_{DSATp}}{k_n V_{DSATn}} \\ &V_{IH} = V_M - \frac{V_M}{g} \qquad V_{IL} = V_M + \frac{V_{DD} - V_M}{g} \\ &\text{with} \quad g \approx \frac{1+r}{(V_M - V_{Tn} - V_{DSATn}/2)(\lambda_n - \lambda_p)} \\ &t_p = \frac{t_{pHL} + t_{pLH}}{2} = 0.69 C_L \Big(\frac{R_{eqn} + R_{eqp}}{2} \Big) \\ &P_{av} = C_L V_{DD}^2 f \end{aligned}$$

Interconnect

Lumped RC: $t_p = 0.69 RC$ Distributed RC: $t_p = 0.38 RC$ RC-chain:

$$\tau_{N} = \sum_{i=1}^{N} R_{i} \sum_{j=i}^{N} C_{j} = \sum_{i=1}^{N} C_{i} \sum_{j=1}^{i} R_{j}$$

Transmission line reflection:

$$\rho = \frac{V_{refl}}{V_{inc}} = \frac{I_{refl}}{I_{inc}} = \frac{R-Z_0}{R+Z_o}$$

CMOS COMBINATIONAL LOGIC

Transistor Sizing using Logical Effort

$$F = \frac{C_L}{C_{g_1}} = \prod_{i=1}^{N} \frac{f_i}{b_i} \qquad G = \prod_{i=1}^{N} g_i \qquad D = t_{p_0} \sum_{j=1}^{N} \left(p_j + \frac{f_j g_j}{\gamma} \right)$$
$$B = \prod_{i=1}^{N} b_i \qquad H = FGB \qquad D_{min} = t_{p_0} \left(\sum_{j=1}^{N} p_j + \frac{N(\frac{N}{\sqrt{H}})}{\gamma} \right)$$